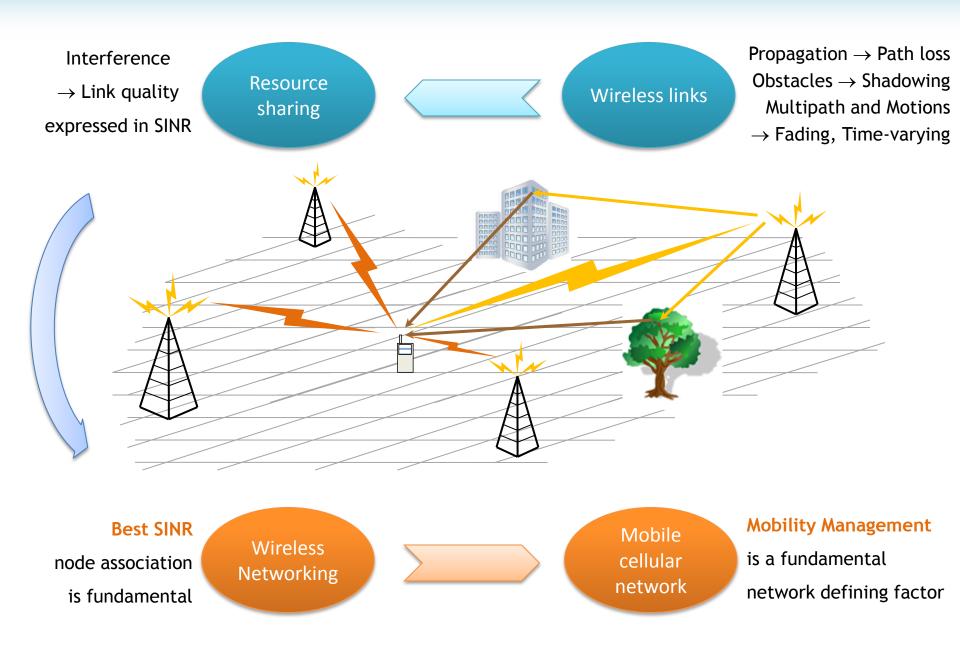
# Wireless Link Quality Modelling and Mobility Management Optimisation for Cellular Networks

PhD Thesis Defence

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Paris, June 20th 2011



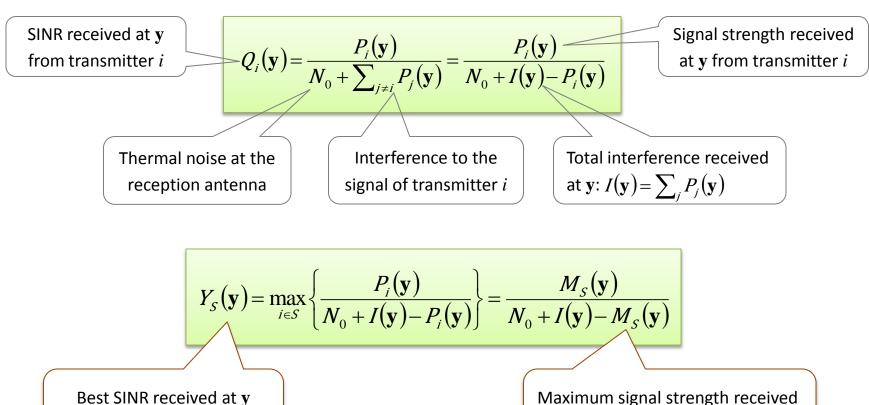
# **Outline of Contributions**

- 1. Wireless Link and Best Signal Quality Modelling
  - 1. Stochastic Geometry Modelling of Wireless Links (IEEE WiOPT 2010)
  - 2. Heavy-Tail Asymptotics of Wireless Links (EURASIP JWCN 2010)
- 2. Level Crossing Analysis of Time-varying Wireless Links
  - 1. Asymptotic Excursions above a Small Level (To be published)
  - 2. Crossings of Successive High Levels (To be published)
- 3. Applications to Mobility Management in Cellular Networks
  - 1. Analytical Model of Handover Measurement with Application to LTE (IEEE ICC 2011)
  - 2. Autonomous Cell Scanning for Small Cell Networks (EURASIP JWCN 2010)
  - 3. Self-optimisation of Neighbour Cell Lists in Macrocellular Networks (IEEE PIMRC'10)

- Presentation of Approach
- Network Assumptions
- Stochastic Geometry Modelling
- Heavy-Tail Asymptotics Modelling



# Approach



from set of transmitters S

Maximum signal strength received at **y** from set *S*:  $M_{S}(\mathbf{y}) = \max_{j \in S} P_{j}(\mathbf{y})$ 

Joint distribution of the total interference I

and the maximum signal strength  $M_s$ 

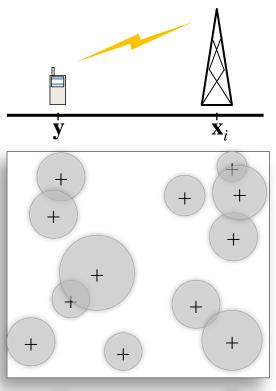
Derive the distribution of the best signal quality  $Y_S$ 

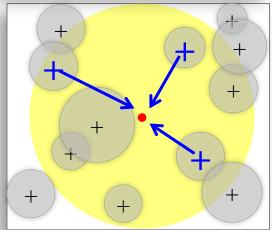
# **Assumptions**

Basic wireless link

$$P_i(\mathbf{y}) = \frac{P_{tx} \cdot Z_i}{l(|\mathbf{y} - \mathbf{x}_i|)} \triangleq \frac{m_i}{l(|\mathbf{y} - \mathbf{x}_i|)}$$

- $\mathbf{y} \in \mathbf{R}^2$  –location of receiver,  $\mathbf{x}_i \in \mathbf{R}^2$  –location of transmitter *i*,
- $P_{tx}$ -node's transmission power,  $\{Z_i\}$  -fading,
- $\{m_i\}$  -virtual Tx power assumed i.i.d. of df  $F_m$ ,  $m := m_1$
- $1/l(r) = r^{-\beta}$  for  $r \in \mathbf{R}_+$  and  $\beta > 2$  –pathloss function
- Interference field as a shot noise
  - $\{\mathbf{x}_i\}$ : Poisson point process with intensity  $\lambda$  on  $\mathbf{R}^2$
  - $\widetilde{\Phi} = \{(\mathbf{x}_i, m_i)\}$ : independently marked Poisson p.p.
  - $\circ \ \ L({\bf x},{\bf y},m)=m/l(|{\bf y}-{\bf x}|)$  : non-negative real resp. function
  - $I(\mathbf{y}) = \sum_{\mathbf{x}_i \in \mathbb{R}^2} P_i(\mathbf{y}) = \sum_{(\mathbf{x}_i, m_i) \in \widetilde{\Phi}} L(\mathbf{y}, \mathbf{x}_i, m_i)$ : SN interference
- Set of observed nodes
  - $B \subset \mathbf{R}^2$ : disk of radius  $R_B$  centred at the receiver,  $\mathbf{y} = \mathbf{0}$
  - *S* = set of nodes uniformly selected from B with prob  $\rho \in [0, 1]$





# Stochastic Geometry Modelling

# **Primary Result**

Joint distribution of I and  $M_S$ 

For  $u \geq 0$  and  $z \in \mathbb{C}$  with a non-negative real part, define:

 $\mathcal{L}_{(I,M_S \le u)}(z) \triangleq \mathbf{E} \{ \mathbf{1}(M_S \le u) \exp(-zI) \}.$ 

Then

$$\mathcal{L}_{(I,M_S \le u)}(z) = \exp\left(-\lambda \int_{\mathbb{R}^2} \left(1 - \mathcal{L}_m(\frac{z}{l(|\mathbf{y} - \mathbf{x}|)})\right) d\mathbf{x} - \rho \lambda \int_B \mathbf{E} \left\{\mathbf{1}\left(\frac{m}{l(|\mathbf{y} - \mathbf{x}|)} > u\right) \exp\left(\frac{-zm}{l(|\mathbf{y} - \mathbf{x}|)}\right) \right\} d\mathbf{x}\right).$$

where  $\mathcal{L}_m(z) \triangleq \mathbf{E}\{\exp(-zm)\}\$  which is the Laplace transform of  $m = m_1$ .

# **Observations**

$$F_{M_S}(u) = \mathcal{L}_{(I,M_S \le u)}(0)$$

Distribution of the Maximum Signal Strength  $M_S$ 

$$F_{M_S}(u) = \exp\left(-\pi\rho\lambda R_B^2\left(1 - F_p(uR_B^\beta)\right) - \pi\rho\lambda \mathbf{E}\{\mathbf{1}(p \le uR_B^\beta)p^\alpha\}u^{-\alpha}\right)$$

where  $\alpha = 2/\beta$ . In particular, if  $B = \mathbb{R}^2$  then:

$$F_{M_S}(u) = \exp\left(-\pi\rho\lambda\mathbf{E}\{p^{\alpha}\}u^{-\alpha}\right)$$

which is a Fréchet dist. with shape  $\alpha$  and scale  $(\pi \rho \lambda \mathbf{E}\{p^{\alpha}\})^{1/\alpha}$ .

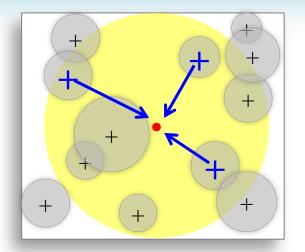
$$\phi_I(w) = \mathcal{L}_{(I,M_S \le +\infty)}(-jw)$$

Characteristic Function of the Total Interference I

Denote 
$$\phi_I(w) = \mathbf{E}\{\exp(jwI)\}\$$
 for  $w \in \mathbb{R}$ . Then

$$\phi_I(w) = \exp\left(-\delta |w|^{\alpha} \left(1 - j\operatorname{sign}(w)\tan(\frac{\pi\alpha}{2})\right)\right)$$

which is the cf of a skewed  $\alpha$ -stable random variable.



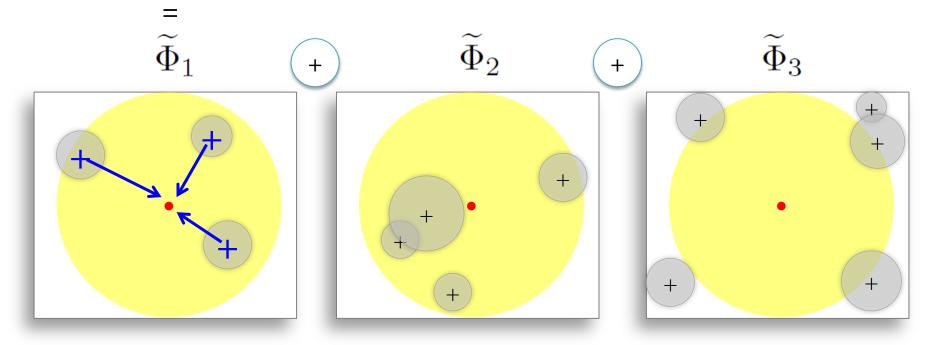
 $\widetilde{\Phi}$ 

# **Skeleton of solution finding**

for  $\mathcal{L}_{(I,M_S \leq u)}(z) \triangleq \mathbf{E} \{ \mathbf{1}(M_S \leq u) \exp(-zI) \}$ 

- Step 1: decompose  $\widetilde{\Phi}$  into three independent independently marked Poisson p.p.
- Step 2: apply the Laplace transform of each shot noise by Prop 2.2.4 in [Baccelli2009]

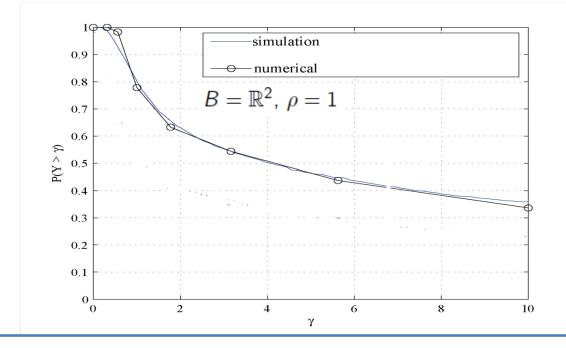
F. Baccelli and B. Blaszczyszyn. "Stochastic Geometry and Wireless Networks, Volume I – Theory". *Foundations and Trends in Networking*, vol. 3(3-4), pp.249-449, 2009.



# Tail Distribution of the Best Signal Quality

For 
$$w \in \mathbb{R}$$
 let  $\phi_{(I,M_S \leq u)}(w) = \mathcal{L}_{(I,M_S \leq u)}(-jw)$ .  
If  $0 < \mathbf{E}\{m^{\alpha}\} < \infty$ , and if  $F_m$  admits a continuous density  $f_m$ :  
 $\overline{F}_{Y_S}(\gamma) = \frac{1}{2\pi} \int_{u=\gamma}^{+\infty} \int_{-\infty}^{+\infty} \phi_{(I,M_S \leq u)}(w) g(w,u) dw du, \forall \gamma > 0$ ,

where 
$$g(w, u) = \exp(-jwu) - (1 + \gamma^{-1}) \exp(jw(1 - \frac{1+\gamma}{\gamma}u))$$
. And  
 $\overline{F}_{Y_S}(0) = 1 - \exp(-\rho\lambda(1 - F_m(0))|B|)$ .



20 June 2011

# Network Assumptions

- Nodes are spatially distributed according to a Poisson point process
- Virtual transmission powers  $\{m_i\}$  are i.i.d. with general distribution  $F_m$
- Unbounded power-law pathloss model,  $1/l(r) = r^{-\beta}$  for  $r \in \mathbf{R}_+$  and  $\beta > 2$

# Main Results

- $\circ\,$  Joint distribution of I and  $M_S$
- Necessary conditions for the integrability & existence of the joint density
- Tail distribution of the best signal quality

# Important Observations

- Total interference is a skewed alpha-stable distribution
- Global maximum signal strength is a Fréchet distribution
- Unbounded power-law pathloss introduces *very heavy-tailed* behaviours of *I* and *M<sub>S</sub>*

independently of the type of fading

# Heavy-Tail Asymptotics

# **Overview**

## Motivation

• Impacts of the pathloss singularity on the tail behaviour of wireless links

## Focus

- Unbounded pathloss:  $1/l(r) = (\max\{r, R_{\min}\})^{-\beta}$  for  $r \in \mathbf{R}_+$ ,  $\beta > 2$ , and  $R_{\min} = 0$
- Bounded pathloss:  $1/l(r) = (\max\{r, R_{\min}\})^{-\beta}$  for  $r \in \mathbf{R}_+$ ,  $\beta > 2$ , and  $R_{\min} > 0$

• Fading  $\{Z_i\}$  are i.i.d. lognormal with parameters  $(0, \sigma_Z)$  with  $0 < \sigma_Z < \infty$ 

- Network area *B* is bounded with radius  $R_B < \infty$ .
- (note: with Poisson p.p. assumption of nodes spatial distribution)

# Roadmap

- Study the tail equivalent distribution of the signal strength  $P_i$
- Asymptotic joint dist of the total interference & max signal strength
- Tail distribution of the best signal quality

# **Tail Behaviour of Signal Strength**

### Theorem

Denote by  $\overline{F}_P$  the tail distribution of the signal strength  $P_i$ . (a) Under unbounded pathloss,  $\overline{F}_P$  is regularly varying:  $\overline{F}_P(x) \sim a^{\alpha} e^{\nu} x^{-\alpha}$ , as  $x \to \infty$ , with  $\alpha = 2/\beta$ . (b) Under bounded pathloss,  $\overline{F}_P$  is rapidly varying:  $\overline{F}_P(x) \sim \text{const} \cdot \frac{\overline{K}_2(x)}{\log x - \mu_2}$ , as  $x \to \infty$ ,

with  $\overline{K}_2(x)$  the tail of a lognormal distribution.

# Interpretation

- The choice of pathloss model has decisive influence on the tail of wireless links
- Decaying power-law path loss is the dominant component
- Under bounded pathloss, the tail of  $P_i$  is determined by the lognormal fading

# **Asymptotic Distribution of Max Signal Strength**

### Theorem

Let  $M_n = \max_{i=1,\mathbf{x}_i \in B}^n P_i$  be the maximum signal strength.

(a) Under unbounded pathloss,  $\exists c_n > 0, d_n \in \mathbb{R}$  such that

$$c_n^{-1}(M_n - d_n) \xrightarrow{d} \Upsilon_{\alpha}(x), \quad as \ n \to \infty,$$

where  $\Upsilon_{\alpha}$  is the std. Fréchet dist. of parameter  $\alpha = 2/\beta$ . (b) Under **bounded** pathloss,  $\exists c_n > 0, d_n \in \mathbb{R}$  such that

$$c_n^{-1}(M_n - d_n) \stackrel{d}{\to} \Lambda, \quad as \ n \to \infty,$$

where  $\Lambda$  is the standard Gumbel distribution.

### Interpretation

• Network densification scenario:  $n \rightarrow \infty$  within a bounded network area *B* 

- Unbounded pathloss: *M<sub>n</sub>* is asymptotically Fréchet distribution under both network extension and network densification
- Bounded pathloss:  $M_n$  is asymp. Gumbel dist. under network densification

# **Asymptotic Joint Distribution**

### Asymptotic Independence

Let  $I_n = \sum_{i=1,\mathbf{x}_i \in B}^n P_i$  be the interference received from n nodes in B. Under the bounded pathloss model, there exists constants  $a_n > 0$  and  $b_n \in \mathbb{R}$ ,  $c_n > 0$  and  $d_n \in \mathbb{R}$  such that

$$\left(\frac{I_n - b_n}{a_n}, \frac{M_n - d_n}{c_n}\right) \stackrel{d}{\to} (\Phi, \Lambda), \quad as \ n \to \infty,$$

and where the coordinates  $\Phi$ , std. Gaussian dist., and  $\Lambda$  are ind.

### Joint Density

Under bounded pathloss, for moderate or large n,  $(M_n, I_n)$ , and  $(M_n, I)$  admit the following approximations

$$\begin{array}{lll} f_{(M_n,I_n)}(u,v) &\approx & f_{M_n}(u) \times f_{I_n}(v), \\ f_{(M_n,I)}(u,v) &\approx & f_{M_n}(u) \times f_I(v), \end{array}$$

where  $f_{M_n}$ ,  $f_{I_n}$ , and  $f_I$  are respectively the pdf of  $M_n$ ,  $I_n$ , and I.

## Tail Distribution of the Best Signal Quality

Let  $Y_n = \max_{i=1,\mathbf{x}_i \in B}^n Q_i$  be the best signal quality received from n nodes uniformly selected from B. Assume the bounded pathloss model and that  $P_{tx}R_{\min}^{-\beta}$  is large. Then the tail dist of  $Y_n$  admits the following approx:  $\overline{F}_{Y_n}(\gamma) \approx \int_{\gamma}^{\infty} \left\{ f_{M_n}(u) \int_{0}^{\infty} \frac{2}{\pi w} e^{-\delta w^{\alpha}} \sin\left(w \frac{u-\gamma}{2\gamma}\right) \right\}$  $\times \cos\left(wu + w\frac{u-\gamma}{2\gamma} - \delta w^{\alpha} \tan\frac{\pi\alpha}{2}\right) \mathrm{d}w \mathrm{d}u.$  $10^{0}$ **Evaluation of**  $E{\xi_n} / E{\xi_{250}}$ Shannon capacity using tail distribution of the best Numerical signal quality Simulation 100 150 50 200 250 0 n [number cells scanned]

# Focus

- Impacts of the singularity of power-law pathloss on wireless links
- Network densification scenario:  $n \rightarrow \infty$  within a bounded network area B
- Fading  $\{Z_i\}$  are i.i.d. lognormal with parameters  $(0, \sigma_Z)$  with  $0 < \sigma_Z < \infty$

# Unbounded pathloss

- Very heavy-tailed behaviours of interference and maximum signal strength
- Interference and maximum signal strength behave *dependently* due the common dominant component corresponding to the pathloss singularity

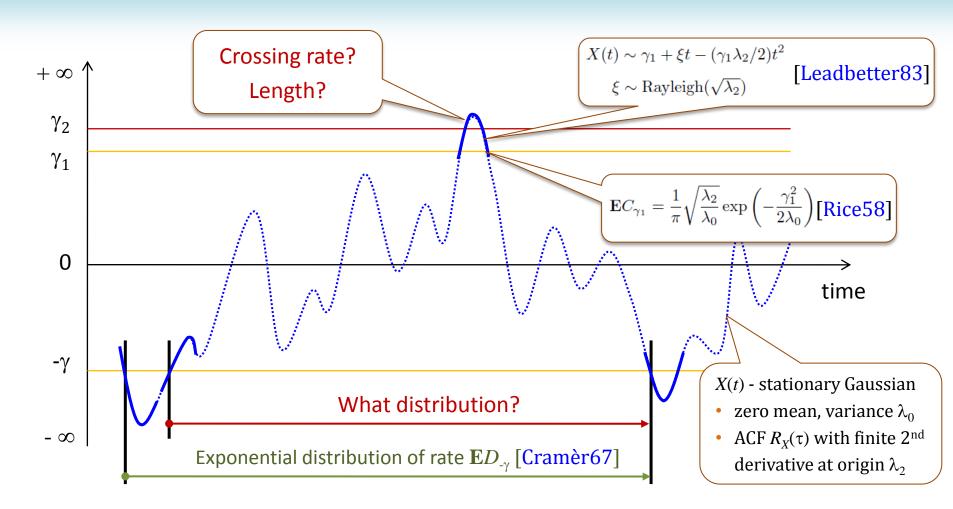
# Bounded pathloss

- Asymptotic ind. between the interference and the max signal strength
- Approximation of the tail distribution of the best signal quality

- Excursions Above a Low Level
- Crossings of Successive High Levels



LEVEL CROSSING PROPERTIES OF A STATIONARY GAUSSIAN PROCESS



[Rice58] S. O. Rice. "Distribution of the duration of fades in radio transmission: Gaussian noise model". *Bell Syst. Tech. J.*, 37(3):581-635, 1958

[Leadbetter83] M. R. Leadbetter, G. Lindgren, and H. Rootzen. *Extremes and Related Properties of Random Sequences and Processes*. Springer Verlag. 1983

[Cramèr67] H. Cramèr and M. R. Leadbetter. *Stationary and related stochastic processes: Sample function properties and their applications,* volume 7. John Wiley and Sons, Inc, 1967.

# Main Result (1/2)

### Excursion Above a Very Low Level

Assume that X(t) admits an autocorrelation function  $R_X(\tau)$  satisfying:

$$R_X(\tau) = 1 - \frac{\lambda_2}{2!}\tau^2 + \frac{\lambda_4}{4!}\tau^4 + o(\tau^4)$$

[Cramèr67]: for the exponential dist of time between two successive down-crossings

with finite  $\lambda_2$  and  $\lambda_4$ , as  $\tau \to 0$ , and satisfying

$$R_X(\tau) = O(\tau^{-a}) \text{ for some } a > 0 \text{ as } \tau \to \infty$$

Then the length  $\tau_u$  of an up-excursion of X(t) above a very small level  $-\gamma \to -\infty$  is asymptotically an exponential dist of rate  $\mu = \mathbf{E}U_{-\gamma}$ , i.e.

 $\mathbf{P}(\tau_u \le \tau) = 1 - e^{-\mu\tau}, \quad as \quad -\gamma \to -\infty.$  By the memorylessness of exponential dist

## Observation

•  $\mathbf{E}U_{-\gamma} \to 0$  as  $-\gamma \to -\infty$ :  $\mathbf{P}(\tau_u \le \tau) \approx 0 \ \forall \tau < \infty$ , i.e. X(t) above a low level most of the time

- Thus, for an excursion above a *low* level, we only know the distribution of length
- By contrast, an excursion above a *high* level is short: length & trajectory by [Leadbetter83]

# Main Result (2/2)

[Leadbetter83]: for the asymp. parabola trajectory of up-excursion above a high lelvel

Crossings of Successive High Levels

Assume that  $R_X(\tau)$  has a finite second derivative at the origin  $\lambda_2$ , and that

 $R_X(\tau) = O(\tau^{-a})$  for some a > 0 as  $\tau \to \infty$ .

Given that X(t) has an up-excursion above level  $\gamma_1$ , then

1. The mean number of up-crossings of X(t) of level  $\gamma_2 \geq \gamma_1$  is

$$\mathbf{E}U_{\gamma_2|\gamma_1} = \exp(-\gamma_1(\gamma_2 - \gamma_1)),$$

2. The length of an up-excursion of X(t) of level  $\gamma_2 \geq \gamma_1$  is

$$\mathbf{P}(T_2 > \tau \mid X(t) \text{ above } \gamma_1) = \exp(-V\tau^2)\exp(-\gamma_1(\gamma_2 - \gamma_1))$$

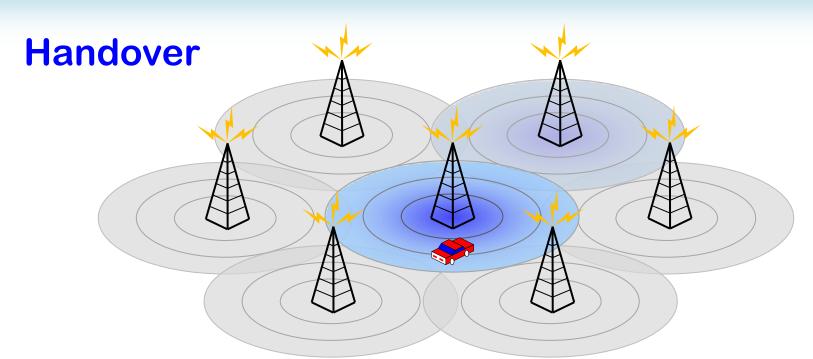
with  $V = \frac{\gamma_1^2 \lambda_2}{8}$  for  $\tau \ge 0$ , and

$$\mathbf{P}(T_2 = 0 \mid X(t) \text{ above } \gamma_1) = 1 - \exp(-\gamma_1(\gamma_2 - \gamma_1))$$

as  $\gamma_1 \to +\infty$ 

- HO Measurement Procedure
- Skeleton of Analytical Solution
- Application to Long Term Evolution





### **Our work**

#### Handover Measurement

- Mobile measures neighbouring cells and reports to the network
- Purpose: to find a suitable HO target when the serving cell's signal deteriorates
- Literature: simulation and parameterspecific approaches, poor in analytical

#### Handover Decision-Execution

- Network decides and executes the connection switching
- Purpose: to perform optimal and reliable connection switching
- Literature: very rich including optimal control, signal prediction, protocol design

# Why is Analytical Model for HO Measurement?

- Strong impacts of HO Measurement on
  - The quality of the handover target
  - The user's experience: service interruptions, throughput degradation
- Complex operation of HO Measurement due to
  - Specific PHY layer procedures: e.g., frame structure, synchronisation,
  - The measurement capability of mobile terminal
  - Combining effect of RRC parameters, e.g. > 10 Triggering Events in WDCMA, 7 in LTE
  - Time-varying and spatial-varying factors, e.g. signal quality, user's mobility
  - The interference nature of a multiple-cell system

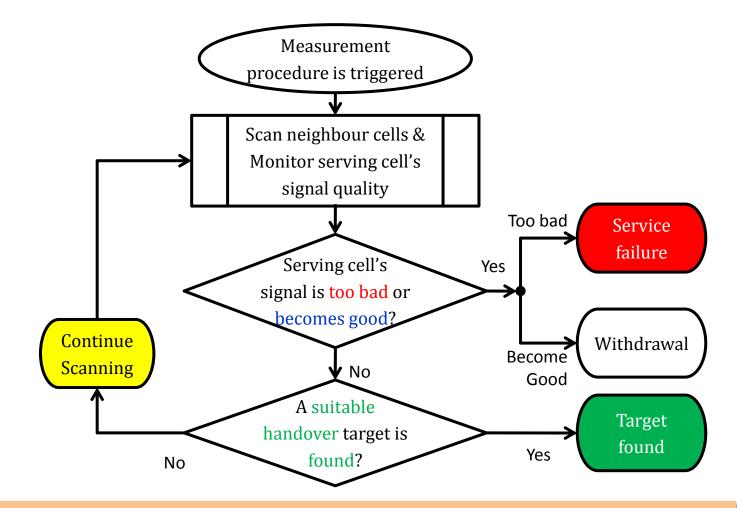
generalised analytical model of handover measurement

is helpful to understand

unified impacts of controlling params + user's mobility + system capabilities

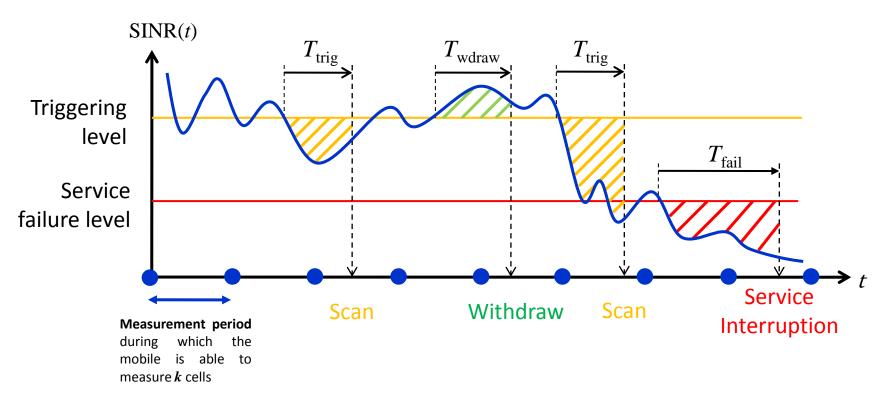
on the system performance

# **Generic HO Measurement Procedure**



A primary objective of the network configuration is to minimise the probability of service failure due to the handover measurement

# **Basic Probabilistic Events**

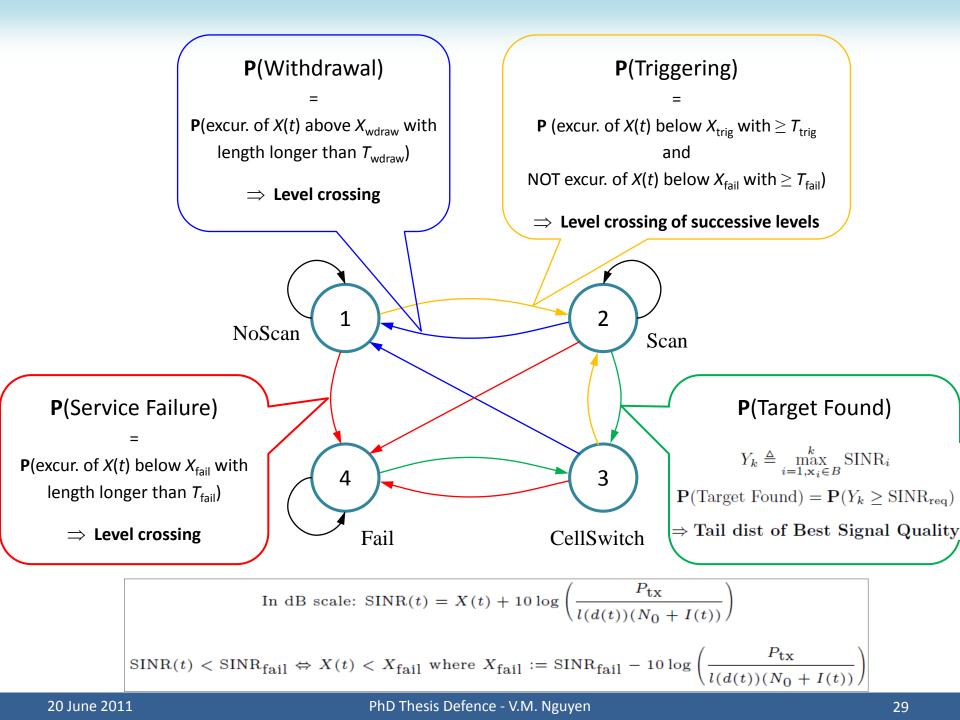


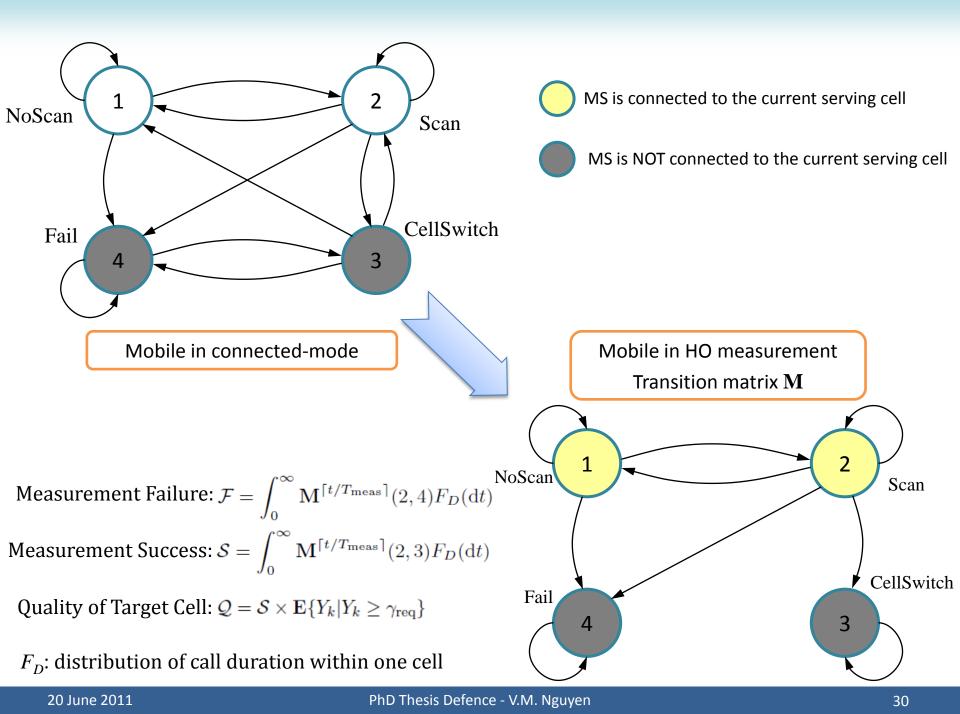
Service Failure  $\equiv$  SINR(t) staying below SINR<sub>fail</sub> during at least  $T_{\text{fail}}$ 

Triggering  $\equiv$  SINR(t) staying below SINR<sub>trig</sub> during at least  $T_{trig}$  and Not Failure

Withdrawal  $\equiv$  SINR(t) staying below SINR<sub>wdraw</sub> during at least  $T_{wdraw}$ 

Target Found  $\equiv$  Best SINR received from measured cells  $\geq$  SINR<sub>req</sub>





# **LTE Handover Measurement**

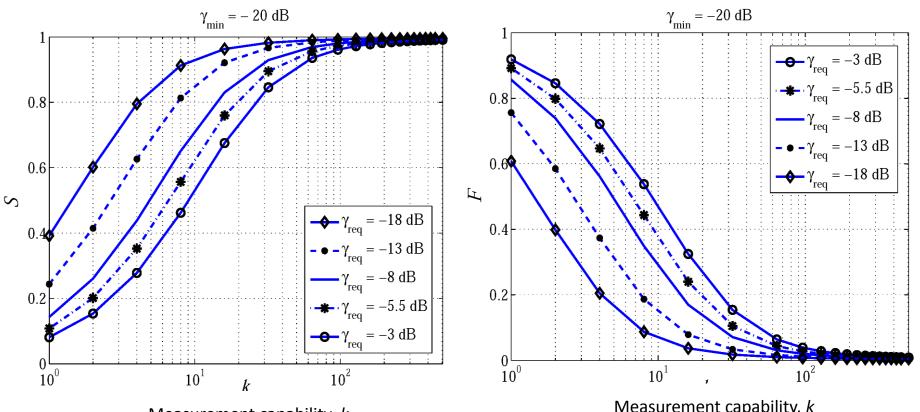
## Intra-freq Measurement

- Intra-frequency measurement is predominant due to frequency-reuse 1 of LTE
- UE measures intra-frequency cells continuously during RRC\_CONNECTED mode
  - $\mathbf{P}(\text{Scanning Triggering}) = 1$
  - $\mathbf{P}(\text{Withdrawal}) = 0$
- UE measures intra-frequency cells autonomously using all 504 physical cell Ids (PCIs)
  - Unlimited candidate set of target cell

Evaluation parameters		
Parameter		Assumption
Scenario	Environment Path loss $(d \text{ in } m)$ User's velocity	Urban macro cell $L(d) = 15.3 + 37.6 \log_{10} d$ $v = 10 \text{ m/s}$
BS	Transmission power Antenna pattern Cell radius	$P_{\rm BS} = 43 \text{ dBm}$ Omnidirectional R = 700  m
Shadowing	Standard deviation Decorr. distance	$\sigma_Z = 10 \text{ dB}$ $d_c = 50 \text{ m}$
Noise	Noise density UE noise figure	= -174  dBm/Hz $N_{\rm F} = 9 \text{ dB}$
Service	Min outage duration Min allowable level	$\tau_{\rm min} = 200 \text{ ms}$ $\gamma_{\rm min} = -20 \text{ to } -5 \text{ dB}$
Measurement	Measurement period Handover margin Required threshold Triggering level Withdrawal level	$T_{\text{meas}} = 200 \text{ ms}$ $\Delta_{\text{HO}} = 2 \text{ dB}$ $\gamma_{\text{req}} = \gamma_{\text{min}} + \Delta_{\text{HO}}$ $\gamma_{\text{t}} = +\infty$ $\gamma_{\text{w}} = +\infty$

#### (a) HO measurement SUCCESS probability

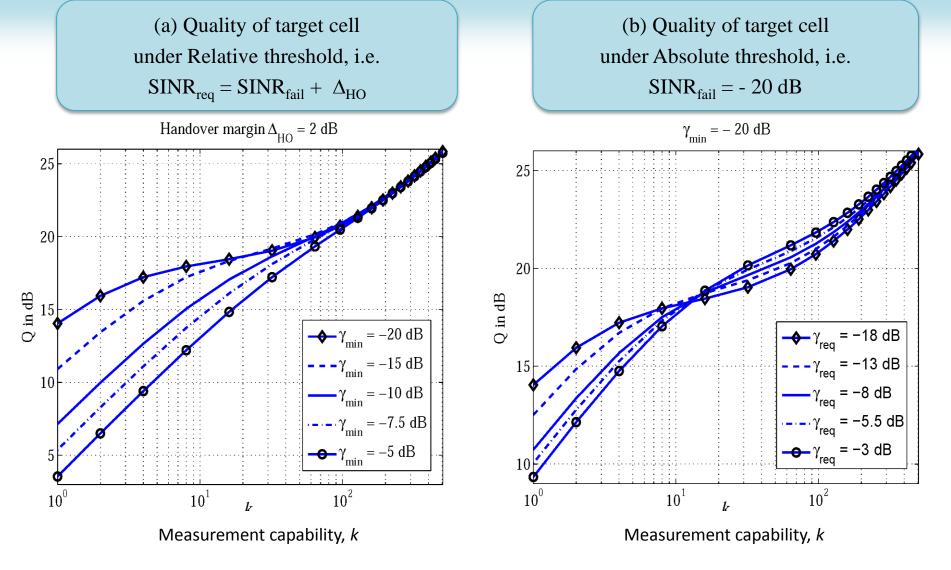
### (b) HO measurement FAILURE probability



Measurement capability, k

Measurement capability, k

- Increasing the measurement capability more than about 10<sup>2</sup> improves the performance very marginally
- Regarding FAILURE probability plot: current LTE requirement for measurement capability of k = 8 seems insufficient for reliable HO performance



- *Relative* threshold: robust service (i.e. low SINR<sub>fail</sub> =  $\gamma_{min}$ ), enhances target cell's quality
- Absolute threshold: crossing point for k in-between 10 and 16. Set low SINR<sub>req</sub> for small k, and set high SINR<sub>req</sub> for big k in order to achieve greater performance

# Analytical model

- Characterise HO measurement procedure as a Markov chain by determining key events associated with a discrete-time model
- Formulate and derive key probabilistic events using the developed results on the best signal quality and on level crossings

## LTE intra-frequency measurement

- High meas. capability enhances the mobility management performance
- Current requirement of 8 intra-freq cells / 200ms seems insufficient
- Measurement capability higher than 10<sup>2</sup> cells / 200ms
  - *Marginal* improvement in the HO measurement performance
  - Significant enhancement of the quality of target cell
- Future applications to Inter-freq and Inter-RAT measurements

### Other applications

- Autonomous scanning for small cell networks
- Self-optimisation of neighbour cell lists

# Conclusion

- Distribution of the Best Signal Quality
  - Method: by means of the joint distribution of interference and max signal strength
  - Stochastic geometry model: exact expression of the tail distribution
  - Heavy-tail asymptotics: an approximation of the tail dsitribution

Level crossing of a stationary Gaussian process

- Length of an excursion above a very low level is exponentially distributed
- Mean number of crossings, length of an excursion of crossings of two successive levels

# Mobility management

- Focus on handover measurement function
- Analytical model using developed results on best signal quality and level crossings
- Application to LTE Intra-frequency measurement



# BACKUP

# **Skeleton of solution finding (cont'd)**

$$\begin{split} \mathbf{E}\{\mathbf{1}(M_{S} \leq u) \exp(-zI)\} \\ &= \\ \mathbf{E}\{\mathbf{1}(M_{S} \leq u) \exp(-z\sum_{i} P_{i}(\mathbf{y}))\} \\ &= \\ \mathbf{E}\{\mathbf{1}(M_{S} \leq u) \exp(-z\sum_{i} P_{i}(\mathbf{y}))\} \\ &= \\ \mathbf{E}\{\exp\left(\sum_{i \in S} (\log \mathbf{1}(P_{i}(\mathbf{y}) \leq u) - zP_{i}(\mathbf{y})) - z\sum_{i \notin S} P_{i}(\mathbf{y}))\} \\ &= \\ \mathbf{E}\{\exp\left(\sum_{i \in S} (\log \mathbf{1}(P_{i}(\mathbf{y}) \leq u) - zP_{i}(\mathbf{y})) - z\sum_{i \notin S} P_{i}(\mathbf{y}))\} \\ &= \\ \mathbf{E}\{\exp\left(\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{1}} \log \mathbf{1}(P_{i}(\mathbf{y}) \leq u) - zP_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{1}} \log \mathbf{1}(P_{i}(\mathbf{y}) \leq u) - zP_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{1}} \log \mathbf{1}(P_{i}(\mathbf{y}) \leq u) - zP_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{2}} P_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{2}} P_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{2}} P_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{3}} P_{i}(\mathbf{y})\right)\} \\ \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{3}} P_{i}(\mathbf{y})\right)\} \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{3}} P_{i}(\mathbf{y})\right)\} \\ \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{3}} P_{i}(\mathbf{y})\right)\} \\ \\ \\ &= \\ \mathbf{E}\{\exp\left(-z\sum_{(\mathbf{x}_{i}, p_{i}) \in \tilde{\Phi}_{3}} P_{i}(\mathbf{y})\right)\} \\ \\ \\ \\ &= \\ \mathbf{$$

## Self-optimisation of Neighbour Cell Lists (IEEE PIMRC 2010)

- Challenge: Manual configuration of NCLs is a big every-day operator's concern
- Solution: Propose measurement-based auto-configuration & self-optimisation
- Conclusion: Attain 99% of scanning success without incurring signalling overhead
- Tools: self-organisation paradigm

### Autonomous Scanning for Small Cell Networks (EURASIP JWCN 2010)

- Challenge: High-density and randomness of small cell networks require for new logic in the implementation of standardised mobility management mechanism
- Solution: Propose and optimise autonomous scanning for max data throughput
- Conclusion: Autonomously scan 30 cells is effective for a common network setting
- Tools: best signal quality for network densification scenario
- Analytical Model of Handover Measurement (IEEE ICC 2011)
  - Challenge: HO Measurement has strong impact on the whole system performance, its operation is complex while its state-of-the-art is weak
  - Solution: Generalised analytical model, then investigation of LTE HO measurement
  - **Conclusion**: Current LTE UE capability seems insufficient for reliable HO performance
  - Tools: best signal quality for network extension & densification, level crossings

# **Service Failure**

#### Formulation

$$SINR = Z + 10 \log \left(\frac{P_{tx}}{l(d)(1+I)}\right)$$

 $\operatorname{SINR}(t) < \operatorname{SINR}_{\operatorname{fail}} \Leftrightarrow Z(t) > Z_{\operatorname{fail}} \text{ where } Z_{\operatorname{fail}} := 10 \log \left( \frac{P_{\operatorname{tx}}}{l(d)(1+I)} \right) - \operatorname{SINR}_{\operatorname{fail}}$ 

Service Failure  $\equiv$  Excursion of Z(t) above  $Z_{\text{fail}}$  with length longer than  $T_{\text{fail}}$ 

## Solution

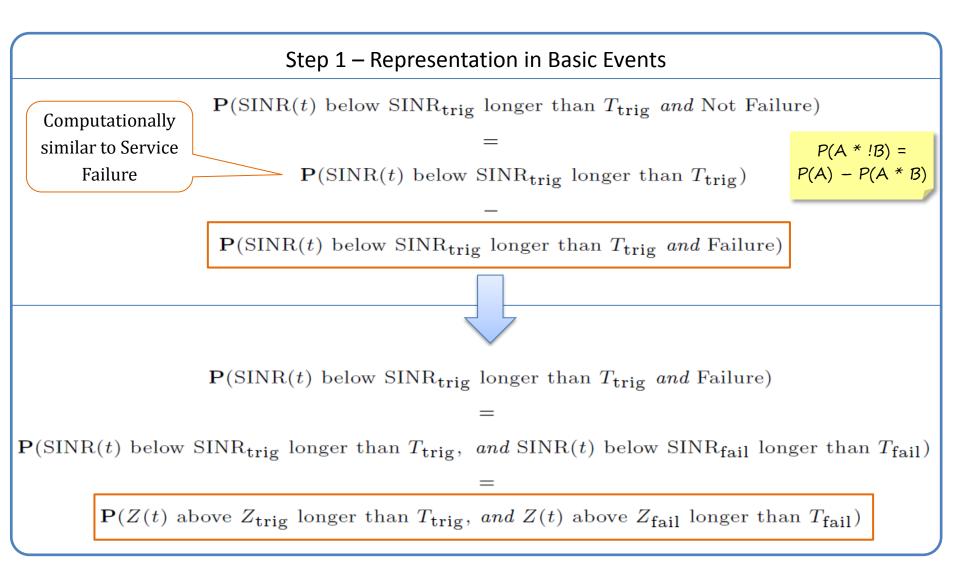
• [Mandayam98]: result for the case of constant  $Z_{\text{fail}}$  assuming constant interference I

• We generalised the result for the case of random  $Z_{\text{fail}}$  taking into account random I

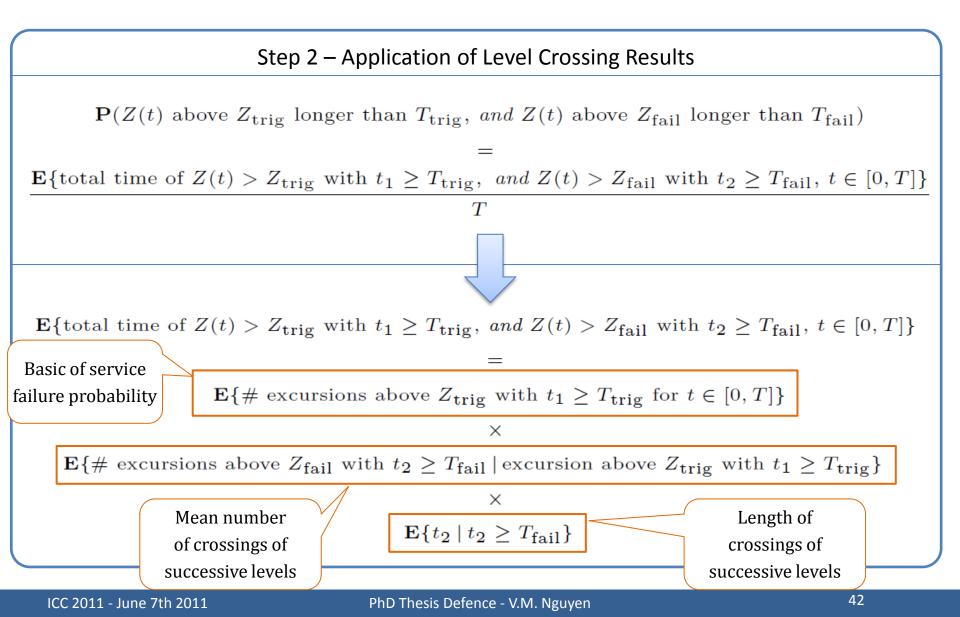
This solution is similarly applied for the probability of scanning withdrawal

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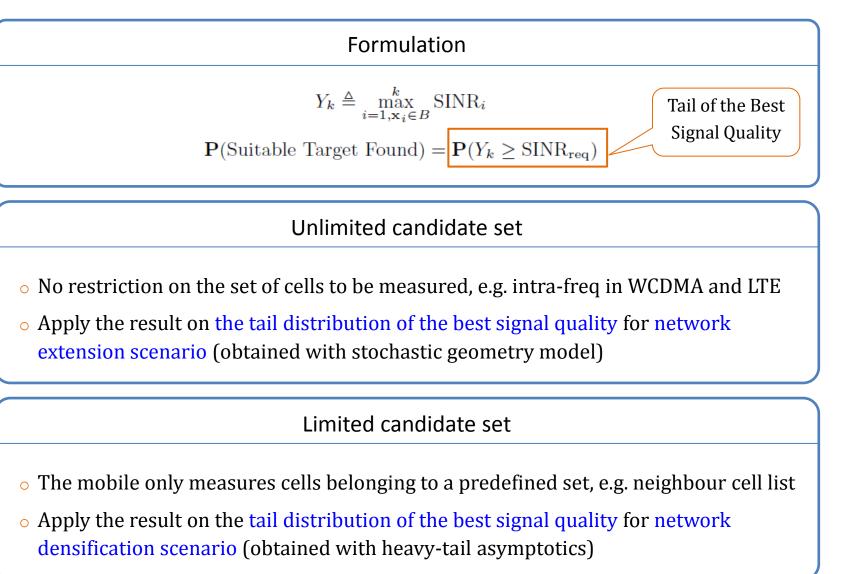
# **Scanning Triggering (1/2)**

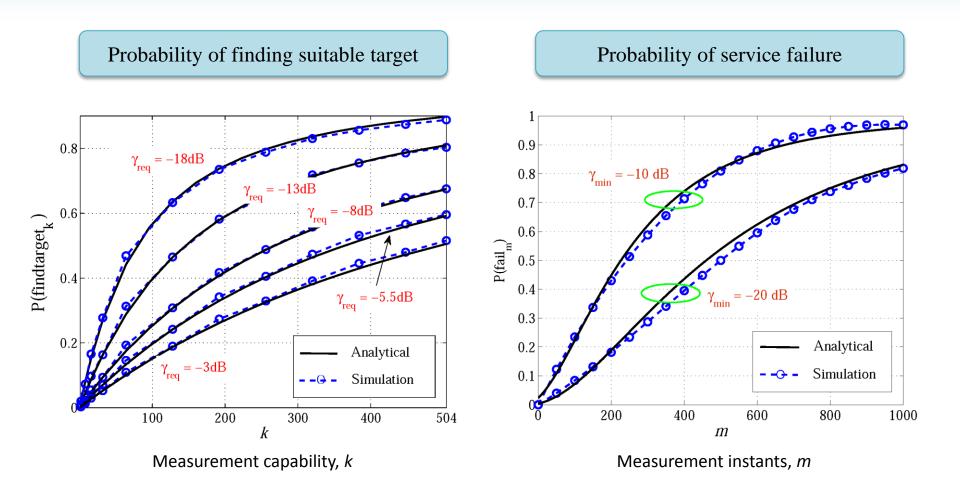


# **Scanning Triggering (2/2)**



# **Suitable Target Found**





- Analytical results agree with simulation
- High *measurement capability* increases the probability of finding a suitable target
- Robust service, i.e. low SINR<sub>fail</sub> (=  $\gamma_{min}$ ), reduces the service failure probability

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